# 3-dimensional Eigenmodal Analysis of Electromagnetic Structures

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# Outline

- Introduction of Femaxx
- Resonant cavities
  - Jacobi-Davidson for linear eigenvalue problems (JDQZ)
- Microwave antennas
  - Jacobi-Davidson for quadratic /nonlinear eigenvalue problems(NLJD)
- Summary

# Introduction of Femaxx

- A large scale software project.
- Developed at ETH Zurich and PSI.
- Simulation tool for electromagnetic structures.
- Parallelized for distributed memory computers.
- Solves electric field vector wave equation.
- Finite element method (FEM).
- 3-dimensional, unstructured tetrahedral mesh.
- Model arbitrary geometry or material property.
- Linear, quadratic and nonlinear eigensolvers.
- Open source software,

http://amas.web.psi.ch/docs/femaxx-doc/html/ .



Swiss Federal Institute of Technology (ETH Zurich)



Paul Scherrer Institute (PSI)

### **Lossless Resonant Cavities**

- Time-harmonic electric field *curl-curl* equation  $\nabla \times (\mu_r^{-1} \nabla \times \boldsymbol{E}(\boldsymbol{x})) - k_0^2 \varepsilon_r \boldsymbol{E}(\boldsymbol{x}) = \boldsymbol{0}, \quad \boldsymbol{x} \in \Omega.$
- Divergence free condition  $\nabla \cdot (\varepsilon_r E(\mathbf{x})) = \mathbf{0}, \quad \mathbf{x} \in \Omega.$
- Perfect electric conductor (PEC) boundary condition  $n \times E(x) = 0$ ,  $x \in \Gamma$ .

 $k_0$ : wavenumber in free space  $\mu_r$ : magnetic relative permeability  $\varepsilon_r$ : electric relative permittivity Ω : computational domain Γ : boundary conductor wall



### **Lossless Resonant Cavities**

- FEM yields a *real-valued linear* eigenvalue problem  $A \mathbf{x} = \lambda M \mathbf{x}, C^T \mathbf{x} = \mathbf{0}, \lambda = k_0^2.$ 
  - A is positive semi-definite:  $a_{ij} = \int_{\Omega} \mu_r^{-1} (\nabla \times N_i) \cdot (\nabla \times N_j) dx$ . *M* is positive definite:  $m_{ij} = \int_{\Omega} \varepsilon_r N_i \cdot N_j dx$ . *C* is a rectangular matrix:  $c_{il} = \int_{\Omega} \varepsilon_r N_i \cdot \nabla N_l dx$ , C = MY. *N<sub>i</sub>* is Nédélec basis function,  $N_l$  is Lagrange basis function, *Y* is a rectangular matrix of  $y_{jl}$ :  $\nabla N_l = \sum_{i=1}^m y_{jl} N_j$ .
- JDSYM eigensolver: Jacobi-Davidson method for real-valued symmetric eigenvalue problem.

### 5-cells Transverse Deflecting Cavity



(a) mesh containing 872'261 elements.



(b) electric field distribution of TM110 mode.





### Cyclotron at PSI

Tetrahedral mesh created from original geometry obtained as STEP file from CAD.



(a) dominant mode: 51.48 MHz



(b) 31st mode: 101.04 MHz





(c) 86th mode: 149.58 MHz



(d) 87th mode: 151.51 MHz

# **Dielectric Lossy Material**

- Complex-valued relative permittivity  $\varepsilon_r = \operatorname{Re}(\varepsilon_r) - i \operatorname{Im}(\varepsilon_r).$
- Perfect electric conductor boundary condition  $n \times E(x) = 0$ ,  $x \in \Gamma$ .

 $(\mu_r, \epsilon_r)$ 

- FEM yields constrained eigenvalue problem  $A \mathbf{x} = \lambda M \mathbf{x}$ ,  $C^T \mathbf{x} = \mathbf{0}$ .
- Entries of M and C depend on  $\varepsilon_r$  $m_{ij} = \int_{\Omega} \varepsilon_r N_i \cdot N_j dx$ ,  $c_{il} = \int_{\Omega} \varepsilon_r N_i \cdot \nabla N_l dx$ .

A is *real* symmetric; M is *complex* symmetric; C is a *complex* rectangular matrix.

# Jacobi-Davidson QZ Eigensolver

### JDQZ solves generalized *non-Hermitian* eigenvalue problem.

- 1: choose start vector  $x_0$ ; it=0.
- when  $it \leq it_{max}$  do 2:
- Solve projected eigenproblem with QZ decomposition. 3:
- 4: Compute an eigenpair closest to target :  $\tau$ .
- 5: if the associated residual is small enough
  - Accept this eigenpair; deflate search space.
- 7: end if

6:

- 8: Restart if necessary.
- Find an approximate solution t of correction equation. 9:
- Project *t* in null space of  $C^T$ . Divergence-free condition 10:
- 11: Orthonormalize t against previous search vectors.
- Expand search space with t. it ++. 12:

13: end

Modified Gram-Schmidt

Krylov solver

Ref: Z. Bai, J. Demmel, J. Dongarra, A. Ruhe and H. van der Vorst. *Templates for the Solution of Algebraic*  $^9$ Eigenvalue Problems. SIAM, Philadelphia, PA, 2000.

## Implementation of Femaxx

- C++ : object-oriented design pattern
- Gmsh : mesh generator
- ParMETIS : mesh partitioner
- Paraview : visualization of electromagnetic field
- Trilinos : parallel objects

# Trilinos in Femaxx

- Epetra
  - parallel multivectors
  - parallel sparse matrices
  - distribution map
  - uses MPI functionality
- AztecOO
  - Iterative solvers for correction equation
- Ifpack
  - Block Jacobi preconditioners

# lfpack

- Block Jacobi preconditioner
  - Extract diagonal blocks by LU and ILU.



Ifpack doesn't support Complex arithmetic.



# Construct double-sized real valued matrix



Ifpack can extract block diagonal elements correctly.
 Reduction of 5% iterations in JDQZ

Ref: David Day, Michael A. Heroux, *Solving Complex-Valued Linear Systems via Equivalent Real Formulations*, SIAM J. Sci. Comput., Vol.23, No.2, pp.480-498.

# lfpack

- Construct proconditioners for Krylov solver
  - Block Jacobi AdditiveSchwarz with LU factorization
    - overlap = 0;
    - partitioner : linear;
    - LU solver : Amesos\_Klu;
  - Block Jacobi AdditiveSchwarz with ILU factorization
    - overlap = 0;
    - partitioner : linear
    - ILU solver : Ifpack\_ILU

Ref: Marzio Sala, Michael Heroux, *Robust Algebraic Preconditioners using IFPACK 3.0*, Sandia Report, SAND2005-0662, 2005.

# Example of dielectric lossy material

- **#Tetrahedra** : 141'149
- Processors : 32
- Quadratic element order
- Number of eigenvalues : 6
- Tolerance : 1.0e-6
- Target : [1.0]
- Krylov solvers : Bi-CGStab
- Preconditioners : Diagonal, Block Jacobi-LU/ILU
- Max iteration of correction equation : 50
- Jmin : 5, Jmax : 10

### Lossy Dielectric Block





Stored energy:  $U = \frac{\varepsilon_0}{2} \int_{\Omega} \operatorname{Re}(\varepsilon_r) |\boldsymbol{E}|^2 d\boldsymbol{x}$ . Dissipated energy:  $U_d = \frac{\varepsilon_0}{2} \int_{\Omega} \operatorname{Im}(\varepsilon_r) |\boldsymbol{E}|^2 d\boldsymbol{x}$ . Resonant frequency:  $f = \frac{\omega}{2\pi}$ . Quality factor:  $Q = \frac{\operatorname{Re}(\lambda)}{2 \times \operatorname{Im}(\lambda)}$ .

Table : Numerical results of the lossy dielectrix block cavity

mode	f (GHz)	Q
1	6.171	11.2
2 3	9.095	5.82
4 5	11.41	7.48
6	11.42	5.49

Ref: S.J. Cooke and B. Levush. In 17th Particle Accelerator Conference, pages 2431–2433, May 1997.

### Lossy Dielectric Block



Mode 1: electric field



Mode 2: electric field



Mode 1: magnetic field



Mode 2: magnetic field

# **Ohmically Lossy Material**

 Considering conduction current density, then timeharmonic electric field vector wave equation is

$$\nabla \times (\mu^{-1} \nabla \times \boldsymbol{E}(\boldsymbol{x})) + i k_0 \sigma Z_0 \boldsymbol{E} - k_0^2 \boldsymbol{\varepsilon}_r \boldsymbol{E}(\boldsymbol{x}) = \boldsymbol{0}, \quad \boldsymbol{x} \in \Omega.$$

σ is ohmic conductivity.  $Z_0 ≈ 377 Ω$  is characteristic impedance in free space.

• FEM yields a constrained *quadratic* eigenproblem  $T(\lambda)x := A x + \lambda R x - \lambda^2 M x = 0, \quad C^T x = 0.$ Entry of  $R: r_{ii} = i \int_{\Omega} \sigma Z_0 N_i \cdot N_i dx.$ 

# **Ohmically Lossy Material**

• Linearize the quadratic eigenvalue problem

$$\begin{bmatrix} A & 0 \\ 0 & I \end{bmatrix} \begin{pmatrix} \mathbf{x}_1 \\ \mathbf{x}_2 \end{pmatrix} = \lambda \begin{bmatrix} -R & M \\ I & 0 \end{bmatrix} \begin{pmatrix} \mathbf{x}_1 \\ \mathbf{x}_2 \end{pmatrix},$$
$$\begin{bmatrix} C & 0 \\ 0 & C \end{bmatrix}^T \begin{pmatrix} \mathbf{x}_1 \\ \mathbf{x}_2 \end{pmatrix} = \mathbf{0}.$$
Identity matrix

Then it is solved by JDQZ eigensolver.

• Dissipated energy now given as:  $U_{d} = \frac{\varepsilon_{0}}{2} \int_{\Omega} \operatorname{Im}(\varepsilon_{r}) |\mathbf{E}|^{2} d\mathbf{x} \longrightarrow U_{d} = \frac{1}{2} \int_{\Omega} \sigma |\mathbf{E}|^{2} d\mathbf{x}$ 

# Nonlinear Jacobi Davidson method (NLJD)

- Algorithm :
  - 1:  $T(\lambda)\mathbf{x}=0$ , choose start vector  $\mathbf{x}_0$ ; it=0.
  - 2: when  $it \leq it_{max}$  do
  - 3: Solve projected nonlinear eigenproblem,  $V_k^T T(\lambda_k) V_k y = 0$ .

Krylov solver

Modified Gram-Schmidt

- 4: Compute a Ritz pair closest to target.
- 5: if the associated residual is small enough
- 6: Return the Ritz pair.
- 7: end if
- 8: Find an approximate solution t of correction equation.
- 9: Project *t* in null space of  $C^T$ . Divergence-free condition
- 10: Orthonormalize *t* against previous search vectors.
- 11: Expand search space with t. it ++.

12: end

### AztecOO

• Correction equation in NLJD:

$$I - \frac{p_{k} u_{k}^{H}}{u_{k}^{H} p_{k}} \left| T(\lambda_{k}) \right| \left| I - \frac{u_{k} u_{k}^{H}}{u_{k}^{H} u_{k}} \right| t_{k} = -r_{k}, \quad t \perp u_{k},$$

where  $\lambda_k$ : approximation of eigenvalue,  $u_k$ : Ritz vector,  $p_k = T'(\lambda_k)u_k$ 

- Solve by AztecOO
  - Epetra\_Operator to calculate projection
  - If pack preconditioner:  $K \approx T(\tau)$ , tau : target

$$\tilde{T} = \left| I - \frac{p_k u_k^H}{u_k^H p_k} \right| T(\tau) \left| I - \frac{u_k u_k^H}{u_k^H u_k} \right|, \quad \tilde{K} = \left| I - \frac{p_k u_k^H}{u_k^H p_k} \right| K \left| I - \frac{u_k u_k^H}{u_k^H u_k} \right|,$$
$$\tilde{K}^{-1} \tilde{T} t_k = -\tilde{K}^{-1} r_k, \quad t \perp u_k,$$

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# Example of half-filled rectangle cavity

- **#Tetrahedra** : 306'001
- Processors : 32
- Number of eigenvalues : 1
- **Tolerance** : 1.0e-6(JDQZ), 1.0e-2(NLJD)
- Target : [100.0]
- Krylov solvers : Bi-CGStab, GMRES(m)
- Preconditioners : Diagonal, Block Jacobi-LU/ILU
- Max iteration of correction equation : 50
- Jmin : 5, Jmax : 10

# Half-filled Rectangle Cavity

**Table**: dominant mode of a rectangle cavity  $(22.86 \times 22.86 \times 10.16 \, mm^3)$ . Half of the cavity is filled by conductor with  $\varepsilon_r = 2.0$  and varying  $\sigma$ .  $\tilde{\omega} = \operatorname{Re}(\sqrt{\lambda})c + i\operatorname{Im}(\sqrt{\lambda})c$ , c: the speed of light

$\sigma$ (S/m)	analytical solution for $\tilde{\omega}/(2\pi)$	$\tilde{\omega}/(2\pi)$ computed by Femaxx
0.1	7.379 + j0.354	7.379 + j0.354
0.5	7.236 + j1.819	7.236 + j1.817
1.0	6.579 + j3.864	6.579 + j3.864
1.3	5.711 + j5.197	5.711 + j5.198





electric field

magnetic field

# Half-filled Rectangle Cavity

- Linear element order
  - JDQZ eigensolver with linearization (dof:664'130)

precondit	outer	inner	average	overall	build time	solver	"speed-
ioner	iteration	iteration		time		time	up"
diag	46	667	14.5	67.51	0.01	49.18	
blkj-ilu	34	348	10.24	64.88	1.17	35.9	1.04
blkj-lu	37	276	7.46	141.39	40.19	85.56	0.48

### NLJD eigensolver, Bi-CGStab (dof:332'065)

preconditi	outer	inner	average	overall	build time	solver	"speed-
oner	iteration	iteration		time		time	up"
diag	18	597	35.12	39.91	0.01	20.92	
blkj-ilu	17	366	22.88	34.72	1.27	16.18	1.15
blkj-lu	30	69	2.46	81.57	40.64	29.92	0.49

# Half-filled Rectangle Cavity

- Quadratic element order (dof:1'852'810)
  - NLJD, Bi-CGStab, 2-level hierarchical basis prec

preconditi	outer	inner	average	overall	build time	solver	"speed-
oner	iteration	iteration		time		time	up"
diag	53	1306	24.64	579.78	6.07	496.7	
blkj-ilu	29	366	12.62	261.00	7.3	177.74	2.22
blkj-lu	23	160	6.96	307.44	46.58	183.83	1.89

### NLJD, GMRES(30), 2-level hierarchical basis prec

preconditi	outer	inner	average	overall	build time	solver	"speed-
oner	iteration	iteration		time		time	up"
diag	44	852	21.85	384.91	0.05	259.91	
blkj-ilu	26	448	19.48	323.64	7.30	233.31	1.19
blkj-lu	19	202	11.88	360.77	46.30	232.04	1.07

## Microwave Antenna

• Bounded computational domain  $\Omega = \Omega_1 \cup \Omega_2 \cup \Omega_3$ .



- $\Omega_1$ : antenna device
- $\Omega_2$ : substrate
- $\Omega_3$ : environment
- $\Gamma$  : absorbing boundary
- $\Omega$ : arbitrary shape, usually sphere
- 1<sup>st</sup> order absorbing boundary condition (ABC):  $\mathbf{n} \times \nabla \times \mathbf{E}(\mathbf{x}) = -i k_0 \sqrt{\mu_r \varepsilon_r} \mathbf{n} \times (\mathbf{n} \times \mathbf{E}(\mathbf{x})), \quad \mathbf{x} \in \Gamma, k_0 = \sqrt{\lambda}.$

### **Microwave Antenna**

• FEM yields a constrained *quadratic* eigenproblem  $A \mathbf{x} + \lambda R \mathbf{x} - \lambda^2 M \mathbf{x} = \mathbf{0}, \quad C^T \mathbf{x} = \mathbf{0}.$ 

$$a_{ij} = \int_{\Omega} \mu_r^{-1} (\nabla \times N_i) \cdot (\nabla \times N_j)$$
  

$$r_{ij} = i \int_{\Gamma} \sqrt{\varepsilon_r / \mu_r} (n \times N_i) \cdot (n \times N_j)$$
  

$$m_{ij} = \int_{\Omega} \varepsilon_r N_i \cdot N_j$$
  

$$c_{il} = \int_{\Omega} \varepsilon_r N_i \cdot \nabla N_l$$
  
integral on boundary

$$\Omega_3$$
 $\Gamma$ 
 $\Omega_1$ 
 $\Omega_2$ 

## Example of microwave antenna

- **#Tetrahedra** : 645'868
- Processors : 32
- Number of eigenvalues : 1
- **Tolerance** : 1.0e-3(JDQZ), 1.0e-3(NLJD)
- Target : [0.1]
- Krylov solvers: Bi-CGStab, GMRES(m)
- Preconditioners : Diagonal, Block Jacobi-LU/ILU
- Max iteration of correction equation : 50
- Jmin : 5, Jmax : 10

## Dielectric Resonant Antenna (DRA)



#### Table: Dominant mode of several DRA.

$\varepsilon_r$	a	b	d	f	$f_{\rm MI}$	Diff.	$Q_1$
	(mm)	(mm)	(mm)	(GHz)	(GHz)	(%)	
79.46	7.45	7.45	2.98	4.644	4.346	6.9	176.8
37.84	8.60	8.60	2.58	6.221	5.934	4.8	31.1
37.84	7.45	7.45	3.51	5.614	5.337	5.2	50.8
20.0	10	10	4	6.545	6.409	2.1	13.3

 $f, Q_1$  and  $Q_2$  are numerical results by using 1st order ABC;  $f_{MI}, Q_{MI}$  are theoretically determined results by using perfect magnetic conductor (PMC) boundary conditions.





Remark:Electric fieldplotted over y-axis: $E_y \approx 0$ , $E_x \approx A k_y \sin(k_y y)$ .

Ref:

1, R. K. Mongia and A. Ittipiboon. IEEE Trans. Antennas Propag., 45(9):1348-1356 (1997).

2, H. Guo, B. Oswald, and P. Arbenz, Opt. Express 20, 5481-5500 (2012).

## Microwave Antenna

- Linear element order
  - JDQZ solver with linearization (dof:162'976)

preconditi	outer	inner	average	overall	build time	solver	"speed-
oner	iteration	illeration		ume		ume	up
diag	20	988	49.78	31.87	0	21.55	
blkj-ilu	18	881	48.94	19.03	0.43	8.93	1.67
blkj-lu	18	259	14.39	22.75	0.3	14.48	1.40

### NLJD solver ,Bi-CGStab(dof:81'488)

pr	econditi	outer	inner	average	overall	build time	solver	"speed-
or	ner	iteration	iteration		time		time	up"
di	ag	19	559	29.42	14.44	0	5.35	
bl	kj-ilu	19	587	30.89	14.60	0.4	5.99	0.99
bl	kj-lu	20	311	15.55	14.53	0.42	7.01	0.99

### Microwave Antenna

- Quadratic element order(dof:439'188)
  - NLJD solver, Bi-CGStab, 2-level hierarchical basis prec

preconditi	outer	inner	average	overall	build time	solver	"speed-
oner	iteration	iteration		time		time	up"
diag	29	745	25.69	47.27	0.04	32.77	
blkj-ilu	17	325	19.12	31.11	0.62	14.31	1.52
blkj-lu	17	223	13.12	36.39	0.81	20.89	1.30

### NLJD, GMRES(30), 2-level hierarchical basis prec

preconditi	outer	inner	average	overall	build time	solver	"speed-
oner	iteration	iteration		time		time	up"
diag	35	952	30.71	77.46	0.02	48.78	
blkj-ilu	20	559	31.06	63.90	1.42	37.30	1.76
blkj-lu	33	1035	35.69	149.32	2.96	117.72	0.67

## Summary

- Femaxx: a parallel 3D finite element eigensolver package.
- Simulation tool for advanced electromagnetic structures.

eigensolver	problem type	applications
JDSYM	Generalized real symmetric eigenvalue problem	lossless resonant cavities
JDQZ	Generalized non-Hermitian eigenvalue problem; Quadratic eigenvalue problem	dielectric lossy material; ohmically lossy material; plasmonic nanostructures
NLJD	Nonlinear eigenvalue problem	ohmically lossy material; plasmonic nanostructures